# Determining the $I^{2} C$ Frequency Divider Ratio for SCL 

by Networking and Multimedia Group<br>Freescale Semiconductor, Inc.<br>Austin, TX

The $\mathrm{I}^{2} \mathrm{C}$ interface is a two-wire, bidirectional serial bus developed by Philips that provides a simple, efficient way to exchange data between integrated circuit (IC) devices. The $I^{2} \mathrm{C}$ interface allows a device to exchange data with other $\mathrm{I}^{2} \mathrm{C}$ devices such as microcontrollers, EEPROMs, real-time clock devices, A/D converters, and LCDs.
This document explains how the frequency divider to calculate the SCL speed of the $\mathrm{I}^{2} \mathrm{C}$ interface is determined for the MPC824x, MPC83xx, MPC85xx, MPC86xx, and P2020 devices. For functional characteristics of the $I^{2} C$ interface of a processor, refer to the device reference manual.

## Contents

1. I2C Signals and Registers ..... 2
2. I2C Frequency Divider Register (I2CFDR/I2CnFDR) 2
3. Determining the Frequency Divider ..... 4
4. Obtaining the Divider Equation ..... 6
5. FDR Divider Tables ..... 10
6. Conclusion ..... 15
7. Revision History ..... 15

## $1 \quad I^{2} C$ Signals and Registers

The $\mathrm{I}^{2} \mathrm{C}$ interface uses serial data (SDA) and serial clock (SCL) signals for data transfer, as shown in Table 1. When the processor is idle or acts as a slave, SCL defaults as an input. The signal patterns driven on SDA represent address, data, or read/write information at different stages of the protocol.

Table 1. SCL and SDA Signals

| Signal Name | Idle State | I/O | Description |
| :---: | :---: | :---: | :--- |
| SCL (serial clock) | HIGH | I | When the processor is idle or acts as a slave, SCL defaults as an input. The unit uses <br> SCL to synchronize incoming data on SDA. The bus is assumed to be busy when SCL is <br> detected low. |
|  |  | O | As a master, the processor drives SCL along with SDA when transmitting. As a slave, the <br> processor drives SCL low for data pacing. |
| SDA (serial data) | HIGH | I | When the processor is idle or in a receiving mode, SDA defaults as an input. The unit <br> receives data from other I ${ }^{2}$ C devices on SDA. The bus is assumed to be busy when SDA <br> is detected low. |
|  |  | O | When transmitting as a master or slave, the processor drives (or transmits) data on SDA <br> synchronous to SCL. |

The $\mathrm{I}^{2} \mathrm{C}$ registers are used for the address, data, configuration, control, and status of the $\mathrm{I}^{2} \mathrm{C}$ controller. These registers are located in either the embedded utilities memory block or the configuration, control, and status block depending on the integrated processor. The $\mathrm{I}^{2} \mathrm{C}$ registers considered in this document are responsible for the digital filter sampling rate and the frequency divider ratio.

## $2 \quad I^{2} \mathrm{C}$ Frequency Divider Register (I2CFDR/I2CnFDR)

The $I^{2} \mathrm{C}$ frequency divider register (I2CFDR) configures the $\mathrm{I}^{2} \mathrm{C}$ sampling and clock rate for the MPC824x, MPC83xx, MPC85xx, and MPC86xx devices. This register is located at 0x0_3004 or 0x004 offset from the base address of one of the following registers, depending on the device:

- Embedded utilities memory block base address register (EUMBBAR for MPC824x devices)
- Configuration, control, and status register base address register (CCSRBAR for MPC85xx and MPC86xx devices)
- Internal memory map registers base address register (IMMRBAR for MPC83xx devices)

Some devices have two $I^{2} \mathrm{C}$ controllers and therefore have I2CnFDR instead of I2CFDR. In this document, unless otherwise stated, I2CFDR refers to both I2CFDR and I2CnFDR. For the integrated processors with two $\mathrm{I}^{2} \mathrm{C}$ controllers, the offsets of the second set of $\mathrm{I}^{2} \mathrm{C}$ registers is always $0 \times 100$ higher than the first set of registers. For example, if a register offset in the first controller is at $0 \times 0 \_3000$, the offset of that register in the second controller is at $0 x 0 \_3100$. Because the offset of I2CFDR in the first controller is 0x0_3004, the offset of the I2CFDR register in the second controller is $0 \times 0 \_3104$. Otherwise, the definition for the second set of registers is exactly the same as that of the first.
The $I^{2} \mathrm{C}$ registers of MPC82xx devices are in the EUMBBAR, which is a PCI configuration register at PCI offset $0 x 78$. The value programmed in this register is the base address for the embedded utilities memory block. The $I^{2}$ C registers of MPC824x, MPC83xx, MPC85xx, and MPC86xx devices are in the CCSRBAR, which is at local offset $0 x 0 \_0000$. The IMMRBAR is at local offset $0 x 0$ for MPC83xx devices. The base
address in both the CCSRBAR and the IMMRBAR is the 12 most-significant bits of the window for configuration accesses including I ${ }^{2} \mathrm{C}$ registers. For some MPC85xx and MPC86xx devices, the base address of the CCSRBAR is the 16 most-significant address bits of the window used for configuration accesses, including $\mathrm{I}^{2} \mathrm{C}$ registers.
The offset of the $\mathrm{I}^{2} \mathrm{C}$ register must be added to that of the EUMBAR, CCSRBAR, or IMMRBAR to obtain the specific address location of that register. For details, refer to the device reference manual.

The bit assignments for the frequency divider ratio (FDR) in the I2CFDR differ among devices, as shown in Table 2, but the definition is the same. The FDR is used to create the SCL speed and to prescale the clock for bit rate selection. The SCL serial bit clock frequency is equal to the <source clock> divided by the divider. The frequency divider value can be changed at any point in a program.

Table 2. FDR Bit and Address Location Summary

| Bit Name <br> (and Assignment) | Register Name | Offset | Device Name/Family |
| :---: | :--- | :--- | :--- |
| FDR (5-0) | $I^{2} \mathrm{C}$ frequency divider register (I2CFDR) | $0 \times 0 \_3004$ | MPC824x |
| FDR (2-7) | $I^{2} \mathrm{C}$ frequency divider register (I2CFDR) | $0 \times 0 \_3004$ | MPC8540/41/60/55E |
| FDR (2-7) | $I^{2}$ Cn frequency divider register (I2CFDR) | $0 \times 0 \_3004,0 \times 3104$ | MPC83xx, MPC85xx where xx is <br> $33 / 36 / 43 / 44 / 45 / 47 / 48 / 67 / 68 / 72[E]$, <br> MPC86xx, and P2020. |

The bit values of the FDR field determine the divider ratio used to create the SCL speed from the source clock of the $\mathrm{I}^{2} \mathrm{C}$ controller. The divider ratio is determined by the value programmed in the FDR as well as by the digital filter sampling rate programmed in the $\mathrm{I}^{2} \mathrm{C}$ controller. The method used to determine this divider is discussed later in this document. The divider for an FDR value is based on the digital filter frequency sampling rate for the part, as follows:

- For MPC824x devices, it is determined by the value programmed in the DFFSR located in the I2CFDR.
- For MPC83xx, MPC85xx and MPC86xx devices, it is determined by the value programmed in the DFSR located in the $I^{2} \mathrm{C}$ digital filter sampling rate register (I2CDFSRR).

The definition of the DFSSR and DFSR bits is the same. To assist in filtering out signal noise, the sample rate is programmable. This field is used to prescale the frequency at which the digital filter takes samples from the $\mathrm{I}^{2} \mathrm{C}$ bus. The resulting sampling rate is the <source clock> divided by the non-zero value set in this field.
Table 3 provides the register information for programming the $\mathrm{I}^{2} \mathrm{C}$ digital filter sampling rate for various integrated processors.

Table 3. Digital Filter Sampling Rate Bit and Address Location Summary

| Bits Name <br> (and <br> Location) | Register Name | Offset | Device Name/Family |
| :--- | :--- | :--- | :--- |
| DFFSR (13-8) | $I^{2} C$ frequency divider register (I2CFDR) | $0 \times 0 \_3004$ | MPC824x |

Table 3. Digital Filter Sampling Rate Bit and Address Location Summary (continued)

| Bits Name <br> (and <br> Location) | Register Name | Offset | Device Name/Family |
| :---: | :--- | :--- | :--- |
| DFSR (2-7) | $I^{2} C$ digital filter sampling rate register (I2CDFSRR) | $0 \times 0 \_3014$ | MPC8540/41/60/55E |
| DFSR (2-7) | $I^{2}$ Cn digital filter sampling rate registers (I2CDFSRR) | $0 \times 0 \_3014$, <br> $0 x 3114$ | MPC83xx, MPC85xx where xx is <br> $33 / 36 / 43 / 44 / 45 / 47 / 48 / 67 / 68 / 72[E]$, <br> MPC86x and P2020. |

Notice that bits 13 through 8 in the DFFSR for the MPC824x devices correspond to bits 2 through 7 in the DFSR for the MPC83xx, MPC85xx, and MPC86xx devices. In this document, DFSR refers to both DFFSR and DFSR because their definitions are the same and they are used the same way to change the $I^{2} \mathrm{C}$ sampling rate of different integrated processors. For details on I2CFDR and I2CDFSRR, consult the device reference manual.

## 3 Determining the Frequency Divider

The bit values of the FDR and DFSR are used to determine the divider ratio between the source clock and the SCL speeds.

### 3.1 Source Clock

The frequency of SCL is based on a source clock divided by the FDR ratio. As Table 4 shows, this source clock is not the same for the all integrated processors. For details, consult the device reference manuals.

Table 4. Source Clock for SCL

| Source Clock | Description | Device Name/Family |
| :---: | :--- | :--- |
| SDRAM clock | The value for the memory clock is based on the peripheral <br> PLL ratio and the input clock for the part <br> (OSC_IN/PCI_SYNC_IN) | MPC824x |
| MPX/CCB <br> clock | The value for this clock is based on the platform PLL ratio <br> and the input clock for the part (SYSCLK/PCI_CLK). | MPC8540/41/60/55E <br> MPC8610 |
| Half of <br> MPX/CCB <br> clock | The value for this clock is based on half of the platform PLL <br> ratio and the input clock for the part (SYSCLK/PCI_CLK). | MPC85xx where xx is <br> $36 / 43 / 45 / 47 / 48 / 67 / 68 / 72[E]$, MPC8641(D), and <br> P2020. |
| Other CCB <br> Clock ratio | The value for this clock is based on half of the platform PLL <br> ratio and the input clock for the part, which is divided by the <br> security PLL ratio. | MPC8533/44-the ratio can be 2:1 or 3:1 (default) <br> for I2C source clock: CCB clock depending on the <br> CCB frequency (refer to the MPC8544 |
| PowerQUICC ${ }^{\text {TM III Integrated Processor }}$ |  |  |
| Hardware Specifications for more details) |  |  |

### 3.2 DFSR

The 6-bit field of the DFSR represents the sampling rate of the digital filter, which factors into the frequency of the SCL because each time the SCL clock switches, the digital sampling logic has an extra period due to bit settings in the FDR. Therefore, the value of DFSR is part of the calculation of the SCL speed. The supported values for sampling rates in the DFSR are $0 \times 01-0 \times 3 F$ or decimal 1 to 63 . If $0 x 00$ is written to the DFSR, for some devices, the register returns to a default value of 0x10. Please refer to the reference manual of the appropriate device to determine the reset value of the DFSR.

### 3.3 FDR Bits and Sub-Dividers

FDR bit assignments vary among devices. Bits 5-0 for MPC824x devices correspond to bits 2-7 of MPC83xx, MPC85xx, and MPC86xx devices. The bit ordering for MPC824x devices is little-endian where bit 0 is the least-significant, and the bit ordering on MPC83xx, MPC85xx, and MPC86xx devices is big-endian where bit 7 is the least-significant bit. For calculating the FDR divider, the 6-bit field of the FDR is partitioned into two groups (A and B). For MPC824x devices, group A consists of bits 5, 1, and 0 and group B consists of bits 4-2. For MPC83xx, MPC85xx and MPC86xx devices, group A consists of bits 2, 6, and 7 and group B consists of bits 3-5. Each group has its own set of dividers based on its bit pattern, as shown in Table 5.

Table 5. FDR Bit Patterns for MPC8xxx Devices

| Bit Values | FDR Bit Locations | Divider |
| :---: | :---: | :---: |
| 000 | Group A <br> Bits 5, 1, 0 for MPC824x devices <br> Bits 2, 6, 7 for MPC83xx, MPC85xx, and MPC86xx devices | 18 |
|  | Group B <br> Bits 4, 3, 2 for MPC824x devices <br> Bits 3, 4, 5 for MPC83xx, MPC85xx, and MPC86xx devices | 16 |
| 001 | Group A <br> Bits 5, 1, 0 for MPC824x devices <br> Bits 2, 6, 7 for MPC83xx, MPC85xx, and MPC86xx devices | 20 |
|  | Group B <br> Bits 4, 3, 2 for MPC824x devices <br> Bits 3, 4, 5 for MPC83xx, MPC85xx, and MPC86xx devices | 32 |
| 010 | Group A <br> Bits 5, 1, 0 for MPC824x devices <br> Bits 2, 6, 7 for MPC83xx, MPC85xx, and MPC86xx devices | 24 |
|  | Group B <br> Bits 4, 3, 2 for MPC824x devices; <br> its 3, 4,5 for MPC83xx, MPC85xx, and MPC86xx devices | 64 |
| 011 | Group A <br> Bits 5, 1, 0 for MPC824x devices <br> Bits 2, 6, 7 for MPC83xx, MPC85xx, and MPC86xx devices | 30 |
|  | Group B <br> Bits 4, 3, 2 for MPC824x devices <br> Bits 3, 4, 5 for MPC83xx, MPC85xx, and MPC86xx devices | 128 |

Determining the I2C Frequency Divider Ratio for SCL, Rev. 5

Table 5. FDR Bit Patterns for MPC8xxx Devices (continued)

| Bit Values | FDR Bit Locations | Divider |
| :---: | :---: | :---: |
| 100 | Group A <br> Bits 5, 1, 0 for MPC824x devices <br> Bits 2, 6, 7 for MPC83xx, MPC85xx, and MPC86xx devices | 10 |
|  | Group B <br> Bits 4, 3, 2 for MPC824x devices <br> Bits 3, 4, 5 for MPC83xx, MPC85xx, and MPC86xx devices | 256 |
| 101 | Group A <br> Bits 5, 1, 0 for MPC824x devices <br> Bits 2, 6, 7 for MPC83xx, MPC85xx, and MPC86xx devices | 12 |
|  | Group B <br> Bits 4, 3, 2 for MPC824x devices <br> Bits 3, 4, 5 for MPC83xx, MPC85xx, and MPC86xx devices | 512 |
| 110 | Group A <br> Bits 5, 1, 0 for MPC824x devices <br> Bits 2, 6, 7 for MPC83xx, MPC85xx, and MPC86xx devices | 14 |
|  | Group B <br> Bits 4, 3, 2 for MPC824x devices <br> Bits 3, 4, 5 for MPC83xx, MPC85xx, and MPC86xx devices | 1024 |
| 111 | Group A <br> Bits 5, 1, 0 for MPC824x devices <br> Bits 2, 6, 7 for MPC83xx, MPC85xx, and MPC86xx devices | 16 |
|  | Group B <br> Bits 4, 3, 2 for MPC824x devices <br> Bits 3, 4, 5 for MPC83xx, MPC85xx, and MPC86xx devices | 2048 |

For example, if the value written in the FDR is $0 x 0 \mathrm{C}(0 \mathrm{~b} 001100)$, the bit patterns assigned are 000 to group A and 011 to group B. As Table 5 indicates, these values correspond to a divider of 18 for group A and 128 for group B. The dividers of groups A and B are used in the calculations to determine the final FDR divider.

## 4 Obtaining the Divider Equation

To simplify the equations for calculating the final divider ratio of source clock to SCL speed, the following variables are used to refer to the FDR bits:

- Variable A represents the divider obtained through group A of Table 5.
- Variable B represents the divider obtained through group B of Table 5.
- Variable C represents the decimal equivalent of the value in the DFSR.

The steps to obtain the divider for calculating the SCL are as follows:

1. Take the decimal value of C; multiply it by 3 ; then divide the result by B.
2. Round the answer to step 1 down to the nearest whole number, that is, 1.x should be rounded down to 1,2 .x should be rounded down to 2 , and so on.
3. Take the result of step 2 and multiply it by 2 .
4. Add the value of A to the results of step 3.
5. Multiply the value of B by the results of step 4 to obtain the new divider.

The above steps can be summarized in the following equation:

$$
\begin{equation*}
\text { Frequency divider }=B \times[A+(\text { floor }(3 \times C \div B) \times 2)] \tag{Eqn. 1}
\end{equation*}
$$

This equation, along with the source clock frequency, is needed to calculate the SCL frequency where:

```
SCL Frequency \(=\) Source clock frequency \(\div\) Frequency divider
```

Eqn. 2
NOTE
In order for Equation 1 to work, the conditions described in Section 4.1, "Required Conditions for the Divider Equation," must be met.

### 4.1 Required Conditions for the Divider Equation

Two conditions must be met for the $\mathrm{I}^{2} \mathrm{C}$ SCL frequency to match the frequency calculated by the divider equation obtained in Section 4, "Obtaining the Divider Equation." The following variables are used in stating the conditions:

- Variable T represents the period of the $\mathrm{I}^{2} \mathrm{C}$ source clock in nanoseconds.
- Variable $\mathrm{t}_{\mathrm{I2} 2 \mathrm{CRM}}$ represents the measured rise time in nanoseconds of the SCL signal from $10 \%$ to $70 \%$ of $\mathrm{V}_{\mathrm{CC}}$.
- Variable B represents the divider obtained through group B of Table 5 as explained in Section 4, "Obtaining the Divider Equation."
- Variable C represents the decimal equivalent of the value in the DFSR as explained in Section 4, "Obtaining the Divider Equation."

The two conditions can be stated as follows:

- Condition 1: $\mathrm{C} \leq 50 \div \mathrm{T}$.

This means that the product of the decimal equivalent of the value in the DFSR and the source clock period must not exceed 50 ns . Thus, given a specific source clock period, the value in the DFSR must not be so high that it violates this condition.

- Condition 2: $\mathrm{B} \times \mathrm{T} \geq \mathrm{t}_{\mathrm{I} 2 \mathrm{CRM}}+3 \times \mathrm{C} \times \mathrm{T}$.

Given a suitable value of DFSR, chosen to satisfy Condition 1, this inequality must also be met to guarantee that the SCL frequency matches the SCL frequency calculated by the divider equation. It is important to note that $\mathrm{t}_{\text {I2CRM }}$ is the measured rise time of the SCL signal, which is defined as the time for the signal to rise from $10 \%$ to $70 \%$ of $\mathrm{V}_{\mathrm{CC}}$.

## NOTE

Note that the rise time must not exceed 300 nanoseconds and that the above two conditions must both be satisfied to ensure that the actual SCL frequency values align with the calculated values. By meeting these conditions, the measured SCL frequency will match the calculated frequency to within 5 kHz . Ignoring either of these conditions may result in larger discrepancies between these frequency values.

### 4.2 Case Studies

To determine the correct values of FDR and DFSR to be used, the following steps should be considered:

1. Choose C (value for DFSR) so that Condition 1 from Section 4.1, "Required Conditions for the Divider Equation" is met, where $\mathrm{C} \leq 50 \div \mathrm{T}$.
2. Choose B (Group B bits of FDR) so that Condition 2 from Section 4.1, "Required Conditions for the Divider Equation" is met, where $\mathrm{B} \times \mathrm{T} \geq \mathrm{t}_{\mathrm{I} 2 \mathrm{CRM}}+3 \times \mathrm{C} \times \mathrm{T}$.
3. Evaluate A (Group A bits of FDR) so that Eqn. 1 and Eqn. 2 from Section 4, "Obtaining the Divider Equation" are met so that:
$\mathrm{A}=[$ Source clock frequency $\div(\mathrm{B} \times$ Desired SCL Frequency $)]-[2 \times$ floor $(3 \times \mathrm{C} \div \mathrm{B})]$.
4. Using the values of $\mathrm{A}, \mathrm{B}$, and C , determine the expected SCL frequency.

### 4.2.1 Example 1 (854x Device)

Consider the case of an 854 x device where the desired SCL frequency is 400 kHz and the source clock frequency is 200 MHz . Hence, the desired frequency divider is 500 .

From Step 1, in Section 4.2, "Case Studies," it can be seen that,

- For Condition $1, \mathrm{C} \leq 50 \div \mathrm{T}$, since the input source clock is 200 MHz , the clock cycle, T , is equal to 5 ns . Therefore, C must be $\leq 10$. For this example, let $\mathrm{C}=8$.

From Step 2, in Section 4.2, "Case Studies," it can be seen that,

- In order for Condition 2 to be met, the rise time of the clock seen at SCL needs to be measured from $10 \%$ to $70 \%$ of $\mathrm{V}_{\mathrm{CC}}$. Since the load and voltage of the system is not expected to change, the rise time measured at this point is expected to be the same rise time when the values of the DFSR and FDR bits are changed later on. Assuming the measured rise time is 120 ns , the value of B that should work can be calculated based on the equation and the value chosen for C , which is 8 .
- Rewriting Condition 2 with the known values, we have $\mathrm{B} \times 5 \geq 120+3 \times 8 \times 5$. Therefore, B must be chosen so that it is $\geq 48$. In Table 5 , it can be seen that 64 is the smallest Group B number that is greater than or equal to 48 . Therefore, we choose B as 64 .

From Step 3, in Section 4.2, "Case Studies," it can be seen that,

- Replacing the known values into A where,

$$
A=[\text { Source clock frequency } \div(B \times \text { SCL Frequency })]-[2 \times \text { floor }(3 \times C \div B)]
$$

we have:

$$
A=[500 \div 64]-[\text { floor }(3 \times 8 \div 64) \times 2]=[7.8125]-[0]=7.8125
$$

- That means that A needs to be the closest value to 7.875 in Table 5. In the table, the closest Group A value to 7.875 is 10 .

From Step 4, in Section 4.2, "Case Studies," it can be seen that,

- Plugging the Table 5 values of $\mathrm{A}, \mathrm{B}$, and the chosen value of C back, we see that the actual frequency divider obtained is $64 \times[10+($ floor $(3 \times 8 \div 64) \times 2)]$, or 640 .
- Therefore, we can program the FDR and DFSR. With C $=8$, the DFSR is $0 b 001000$. Since the Group B divider value used is 64 , bits 3 , 4 and 5 of the MPC854x device are 0,1 , and 0 , respectively, according to Table 5 . Finally, using Table 5 with the Group A divider of 10 , bits 2, 6 , and 7 of the MPC854x device are 1,0 , and 0 , respectively. Hence, the value for the FDR is 0b101000.
- With FDR $=0 \mathrm{~b} 101000$ and $\mathrm{DFSR}=0 \mathrm{~b} 001000$, the expected actual frequency of SCL is 200 $\mathrm{MHz} \div 640$, or $\sim 312.5 \mathrm{kHz}$.


### 4.2.2 Example 2 (8610 Device)

Consider the case of an 8610 device where the desired SCL frequency is 200 kHz and the source clock frequency is 533 MHz . Hence, the desired frequency divider is 2665 .

From Step 1 in Section 4.2, "Case Studies," it can be seen that,

- For Condition $1, \mathrm{C} \leq 50 \div \mathrm{T}$, since the input source clock is 200 MHz , the clock cycle, T , is equal to 1.88 ns . Therefore, C must be $\leq 26$. For this example, let $\mathrm{C}=25$.

From Step 2, in Section 4.2, "Case Studies," it can be seen that,

- In order for Condition 2 to be met, the rise time of the clock seen at SCL needs to be measured from $10 \%$ to $70 \%$ of $\mathrm{V}_{\mathrm{CC}}$. Since the load and voltage of the system is not expected to change, the rise time measured at this point is expected to be the same rise time when the values of the DFSR and FDR bits are changed later on. Assuming the measured rise time is 50 ns , the value of $B$ that should work can be calculated based on the equation and the value chosen for C , which is 25 .
- Rewriting Condition 2 with the known values, we have $\mathrm{B} \times 1.88 \geq 50+3 \times 25 \times 1.88$. Therefore B must be chosen so that it is $\geq 102$. In $\mathrm{T} \alpha \beta \lambda \varepsilon 5,128$ is the smallest Group B number that is greater than or equal to 102 . Therefore, we choose B as 128.

From Step 3, in Section 4.2, "Case Studies," it can be seen that,

- Replacing the known values into A where,

$$
A=[\text { Source clock frequency } \div(B \times \text { SCL Frequency })]-[2 \times \text { floor }(3 \times C \div B)]
$$ we have:

$$
\mathrm{A}=[2665 \div 128]-[\text { floor }(3 \times 25 \div 128) \times 2]=[20.82]-[0]=20.82
$$

- That means that A needs to be the closest value to 20.82 in Table 5 . The closest Group A value to 20.82 is 20.

From Step 4, in Section 4.2, "Case Studies" it can be seen that,

- Plugging the Table 5 values of $\mathrm{A}, \mathrm{B}$, and the chosen value of C back, we see that the actual frequency divider obtained is $128 \times[20+($ floor $(3 \times 25 \div 128) \times 2)]$, or 2560 .
- Therefore, we can program the FDR and DFSR. With $\mathrm{C}=25$, the DFSR is 0 b 011001 . Since the Group B divider value used is 128 , then bits 3,4 and 5 of the MPC8610 device are 0,1 , and 1 , respectively, according to Table 5. Finally, using Table 5 with the Group A divider of 20, bits 2, 6, and 7 of the MPC8610 device are 0,0 , and 1 , respectively. Hence, the value for the FDR is 0b001101.
- With FDR $=0 \mathrm{~b} 001101$ and $\mathrm{DFSR}=0 \mathrm{~b} 011001$, the expected actual frequency of SCL is $533 \mathrm{MHz} \div 2560$, or $\sim 195.3 \mathrm{kHz}$.


## 5 FDR Divider Tables

The divider value for any combination of FDR and DFSR values can be calculated by using the method described in Section 4, "Obtaining the Divider Equation." The supported FDR values are $0 x 00-0 \times 3 F$. The FDR divider table in the reference manual is based on the value DFSR $=0 \mathrm{~b} 10000$, as shown in Table 6.

Table 6. FDR Based on values for DFSR = 0b10000

| FDR | Divider (Decimal) | FDR | Divider (Decimal) |
| :---: | :---: | :---: | :---: |
| 0x00 | 384 | 0x20 | 256 |
| $0 \times 01$ | 416 | $0 \times 21$ | 288 |
| 0x02 | 480 | $0 \times 22$ | 320 |
| $0 \times 03$ | 576 | $0 \times 23$ | 352 |
| 0x04 | 640 | $0 \times 24$ | 384 |
| $0 \times 05$ | 704 | 0x25 | 448 |
| 0x06 | 832 | 0x26 | 512 |
| $0 \times 07$ | 1024 | $0 \times 27$ | 576 |
| $0 \times 08$ | 1152 | $0 \times 28$ | 640 |
| $0 \times 09$ | 1280 | 0x29 | 768 |
| $0 \times 0 \mathrm{~A}$ | 1536 | 0x2A | 896 |
| 0x0B | 1920 | 0x2B | 1024 |
| 0x0C | 2304 | 0x2C | 1280 |
| 0x0D | 2560 | 0x2D | 1536 |
| 0x0E | 3072 | 0x2E | 1792 |
| 0x0F | 3840 | 0x2F | 2048 |
| $0 \times 10$ | 4608 | 0x30 | 2560 |
| $0 \times 11$ | 5120 | $0 \times 31$ | 3072 |
| $0 \times 12$ | 6144 | 0x32 | 3584 |
| $0 \times 13$ | 7680 | $0 \times 33$ | 4096 |
| 0x14 | 9216 | 0x34 | 5120 |

Determining the I2C Frequency Divider Ratio for SCL, Rev. 5

Table 6. FDR Based on values for DFSR = 0b10000 (continued)

| FDR | Divider (Decimal) | FDR | Divider (Decimal) |
| :---: | :---: | :---: | :---: |
| $0 \times 15$ | 10240 | $0 \times 35$ | 6144 |
| $0 \times 16$ | 12288 | $0 \times 36$ | 7168 |
| $0 \times 17$ | 15360 | $0 \times 37$ | 8192 |
| $0 \times 18$ | 18432 | $0 \times 38$ | 10240 |
| $0 \times 19$ | 20480 | $0 \times 39$ | 12288 |
| $0 \times 1 \mathrm{~A}$ | 24576 | $0 \times 3 A$ | 14336 |
| $0 \times 1 B$ | 30720 | $0 \times 3 B$ | 16384 |
| $0 \times 1 \mathrm{C}$ | 36864 | $0 \times 3 C$ | 20480 |
| $0 \times 1 \mathrm{D}$ | 40960 | $0 \times 3 D$ | 24576 |
| $0 \times 1 \mathrm{E}$ | 49152 | $0 \times 3 \mathrm{E}$ | 28672 |
| $0 \times 1 \mathrm{~F}$ | 61440 | $0 \times 3 F$ | 32768 |

Table 7 through Table 9 show the FDR dividers for certain values of the DFSR using the method described in Section 4, "Obtaining the Divider Equation." Table 7 shows the FDR dividers based on a DFSR value of $0 x 01$.

Table 7. FDR Based on a DFSR Value of $0 \times 01$

| FDR | Divider (Decimal) | FDR | Divider (Decimal) |
| :---: | :---: | :---: | :---: |
| $0 \times 00$ | 288 | $0 \times 20$ | 160 |
| $0 \times 01$ | 320 | $0 \times 21$ | 192 |
| $0 \times 02$ | 384 | $0 \times 22$ | 224 |
| $0 \times 03$ | 480 | $0 \times 23$ | 256 |
| $0 \times 04$ | 576 | $0 \times 24$ | 320 |
| $0 \times 05$ | 640 | $0 \times 25$ | 384 |
| $0 \times 06$ | 768 | $0 \times 26$ | 448 |
| $0 \times 07$ | 960 | $0 \times 27$ | 512 |
| $0 \times 08$ | 1152 | $0 \times 28$ | 640 |
| $0 \times 09$ | 1280 | $0 \times 29$ | 768 |
| $0 \times 0 \mathrm{~A}$ | 1536 | $0 \times 2 \mathrm{~A}$ | 896 |
| $0 \times 0 \mathrm{~B}$ | 1920 | $0 \times 2 \mathrm{~B}$ | 1024 |
| $0 \times 0 \mathrm{C}$ | 2304 | $0 \times 2 \mathrm{C}$ | 1280 |
| $0 \times 0 \mathrm{D}$ | 2560 | $0 \times 2 \mathrm{D}$ | 1536 |
| $0 \times 0 \mathrm{E}$ | 3072 | $0 \times 2 \mathrm{E}$ | 1792 |
| $0 \times 0 \mathrm{~F}$ | 3840 | $0 \times 2 F$ | 2048 |

Determining the I2C Frequency Divider Ratio for SCL, Rev. 5

Table 7. FDR Based on a DFSR Value of 0x01 (continued)

| FDR | Divider (Decimal) | FDR | Divider (Decimal) |
| :---: | :---: | :---: | :---: |
| $0 \times 10$ | 4608 | $0 \times 30$ | 2560 |
| $0 \times 11$ | 5120 | $0 \times 31$ | 3072 |
| $0 \times 12$ | 6144 | $0 \times 32$ | 3584 |
| $0 \times 13$ | 7680 | $0 \times 33$ | 4096 |
| $0 \times 14$ | 9216 | $0 \times 34$ | 5120 |
| $0 \times 15$ | 10240 | $0 \times 35$ | 6144 |
| $0 \times 16$ | 12288 | $0 \times 36$ | 7168 |
| $0 \times 17$ | 15360 | $0 \times 37$ | 8192 |
| $0 \times 18$ | 18432 | $0 \times 38$ | 10240 |
| $0 \times 19$ | 20480 | $0 \times 39$ | 12288 |
| $0 \times 1 \mathrm{~A}$ | 24576 | $0 \times 3 \mathrm{~A}$ | 14336 |
| $0 \times 1 \mathrm{~B}$ | 30720 | $0 \times 3 B$ | 16384 |
| $0 \times 1 \mathrm{C}$ | 36864 | $0 \times 3 \mathrm{C}$ | 20480 |
| $0 \times 1 \mathrm{D}$ | 40960 | $0 \times 3 \mathrm{D}$ | 24576 |
| $0 \times 1 \mathrm{E}$ | 49152 | $0 \times 3 \mathrm{E}$ | 28672 |
| $0 \times 1 \mathrm{~F}$ | 61440 | $0 \times 3 F$ | 32768 |

Table 8 shows dividers based on a DFSR value of 0x23. Note the asterisk in the divider in Section 4.2.1, "Example 1 (854x Device)," for an FDR value of 0x03.

Table 8. FDR Based on a DFSR Value of $0 \times 23$

| FDR | Divider (Decimal) | FDR | Divider (Decimal) |
| :---: | :---: | :---: | :---: |
| $0 \times 00$ | 480 | $0 \times 20$ | 352 |
| $0 \times 01$ | 512 | $0 \times 21$ | 384 |
| $0 \times 02$ | 576 | $0 \times 22$ | 416 |
| $0 \times 03$ | $672^{*}$ | $0 \times 23$ | 448 |
| $0 \times 04$ | 768 | $0 \times 24$ | 512 |
| $0 \times 05$ | 832 | $0 \times 25$ | 576 |
| $0 \times 06$ | 960 | $0 \times 26$ | 640 |
| $0 \times 07$ | 1152 | $0 \times 27$ | 704 |
| $0 \times 08$ | 1280 | $0 \times 28$ | 768 |
| $0 \times 09$ | 1408 | $0 \times 29$ | 896 |
| $0 \times 0 \mathrm{~A}$ | 1664 | $0 \times 2 \mathrm{~A}$ | 1024 |
| $0 \times 0 B$ | 2408 | $0 \times 2 \mathrm{~B}$ | 1152 |

Determining the I2C Frequency Divider Ratio for SCL, Rev. 5

Table 8. FDR Based on a DFSR Value of 0x23 (continued)

| FDR | Divider (Decimal) | FDR | Divider (Decimal) |
| :---: | :---: | :---: | :---: |
| 0x0C | 2304 | 0x2C | 1280 |
| 0x0D | 2560 | 0x2D | 1536 |
| 0x0E | 3072 | 0x2E | 1792 |
| 0x0F | 3840 | 0x2F | 2048 |
| 0x10 | 4608 | 0x30 | 2560 |
| $0 \times 11$ | 5120 | $0 \times 31$ | 3072 |
| 0x12 | 6144 | 0x32 | 3584 |
| $0 \times 13$ | 7680 | $0 \times 33$ | 4096 |
| 0x14 | 9216 | 0x34 | 5120 |
| 0x15 | 10240 | 0x35 | 6144 |
| 0x16 | 12288 | 0x36 | 7168 |
| $0 \times 17$ | 15360 | $0 \times 37$ | 8192 |
| 0x18 | 18432 | 0x38 | 10240 |
| 0x19 | 20480 | 0x39 | 12288 |
| 0x1A | 24576 | 0x3A | 14336 |
| 0x1B | 30720 | 0x3B | 16384 |
| 0x1C | 36864 | 0x3C | 20480 |
| 0x1D | 40960 | 0x3D | 24576 |
| 0x1E | 49152 | 0x3E | 28672 |
| 0x1F | 61440 | 0x3F | 32768 |

Table 9 shows dividers based on a DFSR value of 0x34. Note the asterisk in the divider in Section 4.2.2, "Example 2 ( 8610 Device)," for an FDR value of 0x0E.

Table 9. FDR Based on a DFSR Value of 0x34

| FDR | Divider (Decimal) | FDR | Divider (Decimal) |
| :---: | :---: | :---: | :---: |
| $0 \times 00$ | 576 | $0 \times 20$ | 448 |
| $0 \times 01$ | 608 | $0 \times 21$ | 480 |
| $0 \times 02$ | 672 | $0 \times 22$ | 512 |
| $0 \times 03$ | 768 | $0 \times 23$ | 544 |
| $0 \times 04$ | 832 | $0 \times 24$ | 576 |
| $0 \times 05$ | 896 | $0 \times 25$ | 640 |
| $0 \times 06$ | 1024 | $0 \times 26$ | 704 |
| $0 \times 07$ | 1216 | $0 \times 27$ | 768 |

Determining the I2C Frequency Divider Ratio for SCL, Rev. 5

Table 9. FDR Based on a DFSR Value of 0x34 (continued)

| FDR | Divider (Decimal) | FDR | Divider (Decimal) |
| :---: | :---: | :---: | :---: |
| 0x08 | 1408 | $0 \times 28$ | 896 |
| 0x09 | 1536 | 0x29 | 1024 |
| 0x0A | 1792 | 0x2A | 1152 |
| 0x0B | 2176 | 0x2B | 1280 |
| 0x0C | 2560 | 0x2C | 1536 |
| 0x0D | 2816 | 0x2D | 1792 |
| 0x0E | 3328* | 0x2E | 2048 |
| 0x0F | 4096 | 0x2F | 2304 |
| 0x10 | 4608 | 0x30 | 2560 |
| $0 \times 11$ | 5120 | 0x31 | 3072 |
| $0 \times 12$ | 6144 | 0x32 | 3584 |
| $0 \times 13$ | 7680 | 0x33 | 4096 |
| 0x14 | 9216 | 0x34 | 5120 |
| 0x15 | 10240 | 0x35 | 6144 |
| 0x16 | 12288 | 0x36 | 7168 |
| 0x17 | 15360 | 0x37 | 8192 |
| 0x18 | 18432 | 0x38 | 10240 |
| $0 \times 19$ | 20480 | 0x39 | 12288 |
| $0 \times 1 \mathrm{~A}$ | 24576 | 0x3A | 14336 |
| 0x1B | 30720 | 0x3B | 16384 |
| 0x1C | 36864 | 0x3C | 20480 |
| 0x1D | 40960 | 0x3D | 24576 |
| 0x1E | 49152 | 0x3E | 28672 |
| 0x1F | 61440 | 0x3F | 32768 |

In some cases, the same FDR divider table can be used for various DFSR values using the first three steps of the calculation method in Section 4, "Obtaining the Divider Equation." The constants in the first three steps can be summed up as rounding down the decimal portion of the results of the DFSR (from 0.9 and below) divided by variable $B$ and then multiplying the remaining integer by 6 . Calculations show that the DFSR values with the same integer result when divided by 5.4 (that is, $0.9 \times 6$ ) have the same FDR divider table. Therefore, DFSR values ending in decimal 5 and below have the same integer result. Similarly, DFSR values ending in decimal 6-10 have the same FDR dividers, as well as those ending in decimal 11-15.
DFSR values of $0 \times n 0-0 \times n 5$ have the same FDR divider table, DFSR of $0 \times n 6-0 \times n$ A have the same divider table, and DFSR of $0 \mathrm{x} n \mathrm{~B}-0 \mathrm{x} n \mathrm{~F}$ have the same DFSR. The $n$ must be the same for this shortcut to work.

An exception occurs when $0 x 00$ is written to DFSR because for some devices it resets to a default value of $0 \times 10$. Therefore, if the FDR divider table for $\mathrm{DFSR}=0 \times 10$ is calculated, the divider values are the same for DFSR $=0 \times 10-0 x 15$. Similarly, if the FDR divider table for DFSR $=0 \times 3 B$, the divider values are the same for DFSR $=0 \times 3 \mathrm{~B}-0 \times 35$. Considering these results, it is clear that Table 7 works for DFSR values of $0 \times 01-0 \times 05$, Table 6 works for DFSR values of $0 \times 10-0 \times 15$, Table 8 works for DFSR values of $0 \times 20-0 \times 25$, and Table 9 works for DFSR values of $0 \times 30-0 \times 35$.

## 6 Conclusion

The speed of SCL can be calculated based on the source clock speed divided by a programmable divider. The divider is calculated on the basis of the entries in the DFSR and FDR bits of the $\mathrm{I}^{2} \mathrm{C}$ controller, if the conditions in Section 4, "Obtaining the Divider Equation," are satisfied. For bit-specific information on DFFSR/DFSR and FDR, consult the reference manuals of the MPC824x, MPC83xx, MPC85xx and MPC86xx devices. The method described Section 3, "Determining the Frequency Divider," along with the conditions in Section 4, "Obtaining the Divider Equation," can be used to obtain the FDR dividers for DFSR cases not already covered in this document.

## NOTE

The required conditions provided in Section 4, "Obtaining the Divider Equation," must be met to accurately calculate the expected divider and SCL frequency (within 5 kHz ). Although this method yields the expected divider used to determine the SCL value, allow for some margin of error depending on the hardware design and the impact of the slave on SCL frequency.

## 7 Revision History

Table 10 provides a revision history for this application note.
Table 10. History Table

| Revision <br> Number | Date | Substantive Change(s) |
| :---: | :---: | :--- |
| 5 | $12 / 2008$ | - Added some equation format information to Section 4, "Obtaining the Divider Equation" <br> - Added a note to Section 4.1, "Required Conditions for the Divider Equation." <br> - Added Section 4.2, "Case Studies" <br> - Changed the FDR value and example details in Section 4.2.1, "Example 1 (854x Device)" and <br> Section 4.2.2, "Example 2 (8610 Device)." |
| 4 | $11 / 2008$ | - In Table 2, Table 3, and Table 4, updated device name/family information." <br> - Updated sentence below Table 5 to reflect the correct FDR value. <br> - Updated the last sentence of the first paragraph and corresponding table in Section 4, "Obtaining <br> the Divider Equation." <br> - Replaced Section 3.4.1 and 3.4.2 with Section 4.2.1 and Section 4.2.2 <br> - Added Section 4, "Obtaining the Divider Equation." <br> - Updated Section 6, "Conclusion," to reflect conditions required prior to calculations. |
| 3 | $01 / 2008$ | - Added text relevant to the MPC8544[E], MPC8568[E], MPC8572[E], MPC8610 and <br> MPC8641(D). |

Table 10. History Table (continued)

| $\begin{array}{c}\text { Revision } \\ \text { Number }\end{array}$ | Date | Substantive Change(s) |
| :---: | :---: | :--- |
| 2 | $04 / 2006$ | $\begin{array}{l}\text { • Updated Table 4, "Source Clock for SCL," to include correct information for source clock of } \\ \text { MPC85xx where xx is 43/45/47/48 [E] }\end{array}$ |
| 1 | $01 / 2006$ | $\begin{array}{l}\text { - Renamed document to "Determining the I }\end{array}$ T Frequency Divider Ratio for SCL." Replaced all |
| content specific to MPC824x with information applicable to MPC824x, MPC83xx, and MPC85xx |  |  |
| devices. Rearranged and added sections and tables. |  |  |$]$

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Japan
0120191014 or
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